

Projecto e Controlo em Lógica Digital

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State Machines

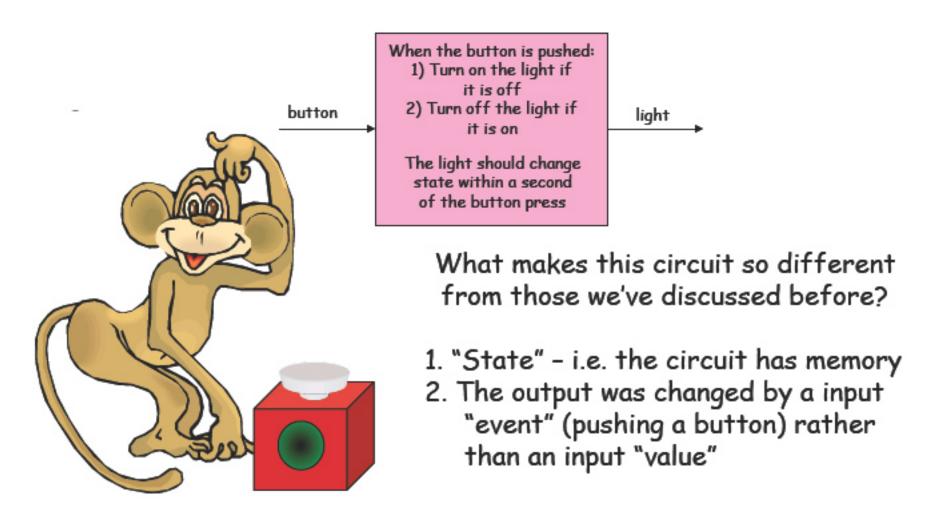
Refs:

Cyclone II device Handbook, Altera corp.
Quartus II Handbook, Altera corp.
DE2 documentation
Verilog HDL, S. Palnitkar, Prentice Hall

Sequential Logic

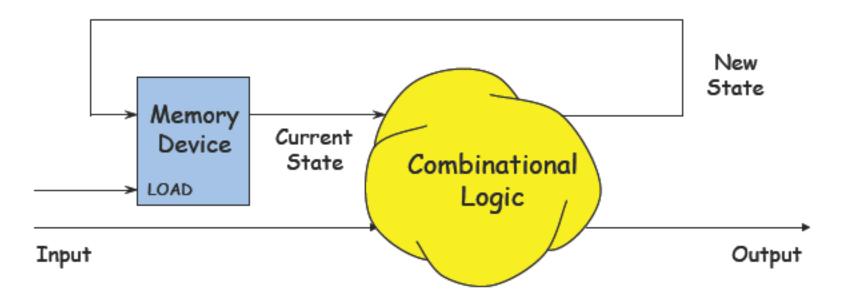
... Something we couldn't build with combinatorial logic

What if you were given the following design specification:



Digital State

One model of what we'd like to build

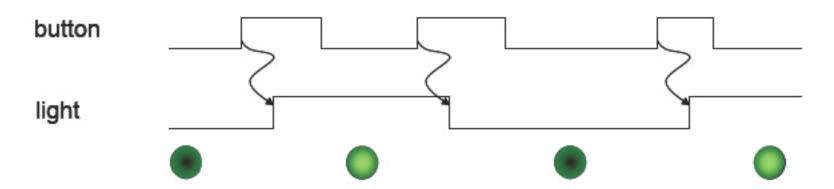


Plan: Build a Sequential Circuit with stored digital STATE -

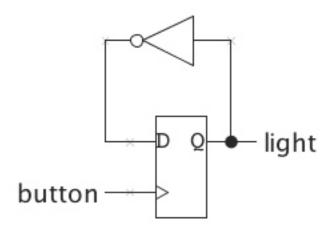
- Memory stores CURRENT state, produced at output
- Combinational Logic computes
 - NEXT state (from input, current state)
 - OUTPUT bit (from input, current state)
- State changes on LOAD control input

When Output depends on input and current state, circuit is called a Mealy machine. If Output depends <u>only</u> on the current state, circuit is called a Moore machine.

Implementation for on/off button



module onoff(input button, output reg light);
 always @(posedge button) light <= ~light;
endmodule</pre>



Synchronous on/off button

When designing a system that accepts many inputs it would be hard to have input changes serve as the system clock (which input would we use?). So we'll use a single clock of some fixed frequency and have the inputs control what state changes happen on rising clock edges.



Resetting to a known state

Usually one can't rely on registers powering-on to a particular initial state*. So most designs have a RESET signal that when asserted initializes all the state to known, mutually consistent initial values.

^{*} Actually, our FPGAs will reset all registers to 0 when the device is programmed. But it's nice to be able to press a reset button to return to a known state rather than starting from scratch by reprogramming the device.



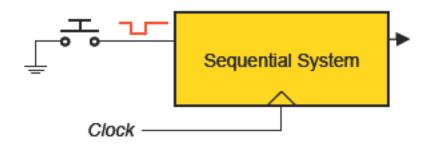
Clocks are fast, we're slow!

The circuit on the last slide toggles the light on every rising clock edge for which button is 1. But clocks are fast (27MHz!) and our fingers are slow, so how do we press the button for just one clock edge? Answer: we can't, but we can can add some state that remembers what button was last clock cycle and then detect the clock cycles when button changes from 0 to 1.



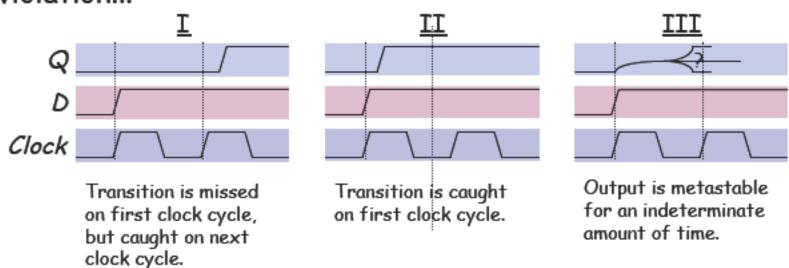
Asynchronous Inputs in Sequential Systems

What about external signals?



Can't guarantee setup and hold times will be met!

When an asynchronous signal causes a setup/hold violation...



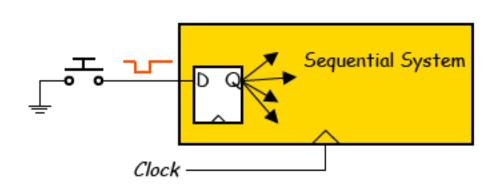
Q: Which cases are problematic?

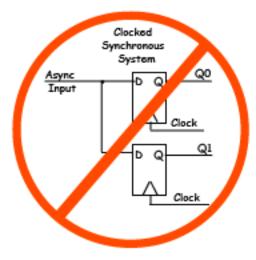


Asynchronous Inputs in Sequential Systems

All of them can be, if more than one happens simultaneously within the same circuit.

Guideline: ensure that external signals directly feed exactly one flip-flop



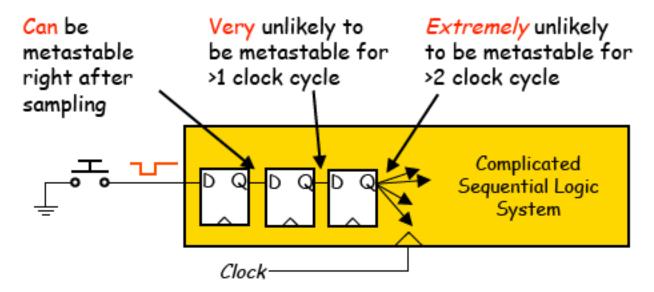


This prevents the possibility of I and II occurring in different places in the circuit, but what about metastability?



Handling Metastability

- Preventing metastability turns out to be an impossible problem
- High gain of digital devices makes it likely that metastable conditions will resolve themselves quickly
- Solution to metastability: allow time for signals to stabilize



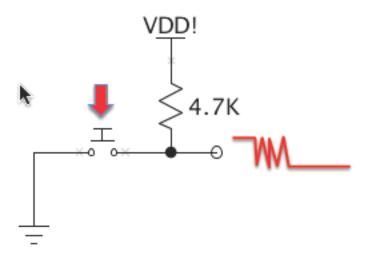
How many registers are necessary?

- Depends on many design parameters (clock speed, device speeds, ...)
- In 6.111, a pair of synchronization registers is sufficient



One last little problem...

Mechanical buttons exhibit contact "bounce" when they change position, leading to multiple output transitions before finally stabilizing in the new position:



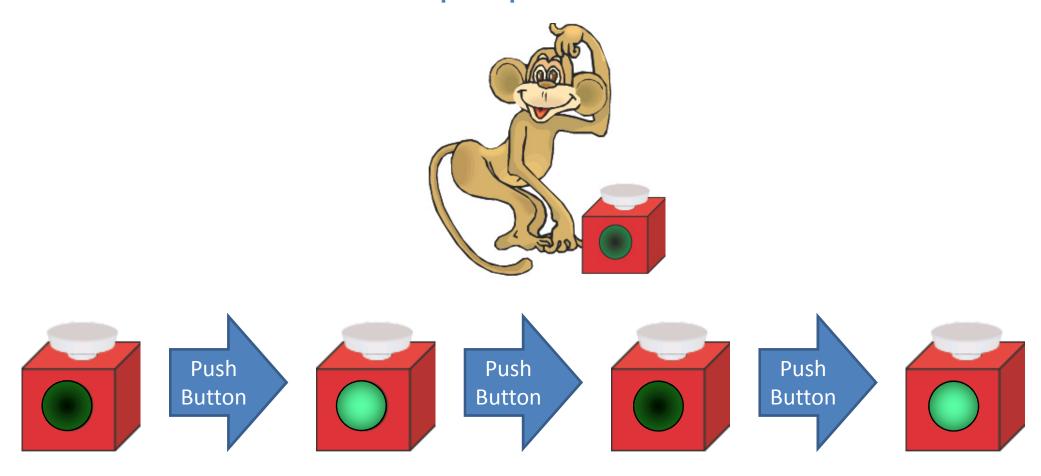
We need a debouncing circuit!

```
// Switch Debounce Module
// use your system clock for the clock input
// to produce a synchronous, debounced output
// DELAY = .01 sec with a 27Mhz clock
module debounce #(parameter DELAY=270000)
                (input reset, clock, noisy,
                 output reg clean);
   reg [18:0] count;
   reg new;
   always @(posedge clock)
     if (reset) // return to known state
       beain
         count \leftarrow 0;
         new <= noisy:
         clean <= noisy:
       end
     else if (noisy != new)
                               // input changed
       beain
         new <= noisy;
         count \leftarrow 0;
       end
     else if (count == DELAY) // stable!
       clean <= new;
     else
                               // waiting...
       count <= count+1;
endmodule
```

On/off button: final answer

```
module onoff_sync(input clk, reset, button_in,
                  output reg light);
  // synchronizer
  req button, btemp;
  always @(posedge clk)
    {button,btemp} <= {btemp,button_in};
  // debounce push button
  wire bpressed;
  debounce db1(.clock(clk),.reset(reset),
               .noisy(button),.clean(bpressed));
  req old_bpressed; // state last clk cycle
  always @ (posedge clk) begin
    if (reset)
      begin light <= 0; old_bpressed <= 0; end</pre>
    else if (old_bpressed==0 && bpressed==1)
      // button changed from 0 to 1
      light <= ~light;
    old_bpressed <= bpressed;
  end
endmodule.
```

A Simple problem...



always @ (posedge button)
begin
 led<=~led;
end</pre>

A Simple problem...



Question #1:

If the button has a value "low" the led should be 'ON" or "OFF"?

Question #2:

If the button has a value "high" the led should be 'ON" or "OFF"?

Question #3:

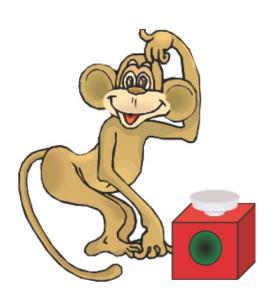
If the button is pressed the led should be 'ON" or "OFF"?

Depends on the STATE!!

This is a State machine!!



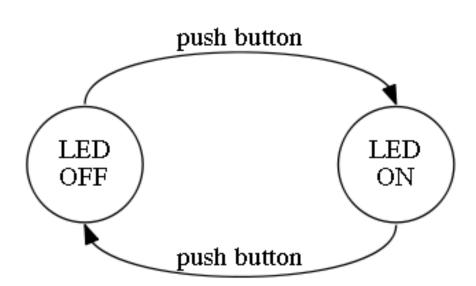
A Simple problem...



Let's implement a state machine for this simple problem

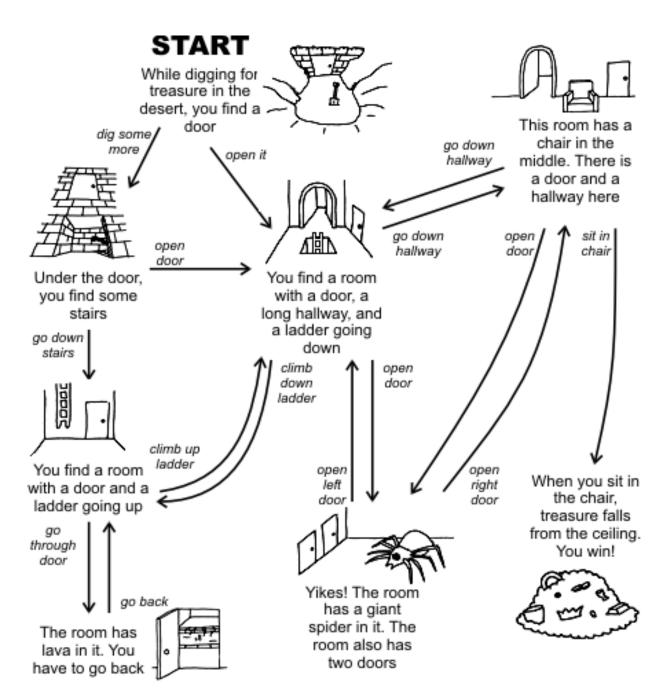


State Diagram:

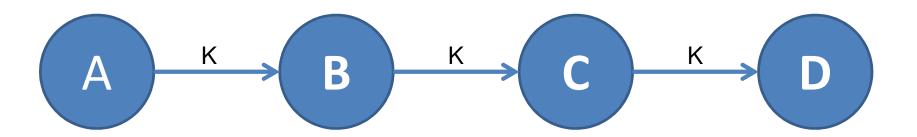


State Machines

Can get complicated...

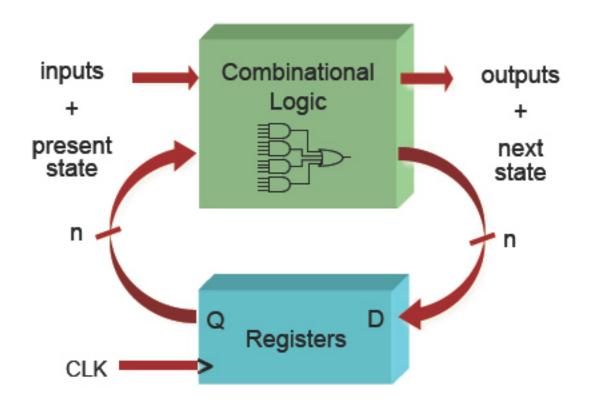


The worst one!!



Finite State Machines

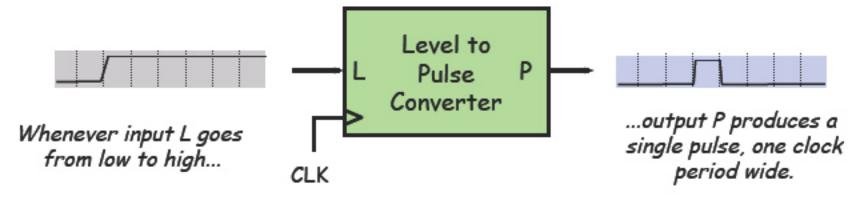
- Finite State Machines (FSMs) are a useful abstraction for sequential circuits with centralized "states" of operation
- At each clock edge, combinational logic computes outputs and next state as a function of inputs and present state



Design Example: Level-to-Pulse

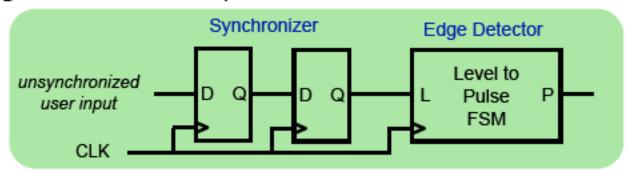
- A level-to-pulse converter produces a single -cycle pulse each time its input goes high.
- It's a synchronous rising-edge detector.
- Sample uses:
 - Buttons and switches pressed by humans for arbitrary periods of time
 - Single-cycle enable signals for counters



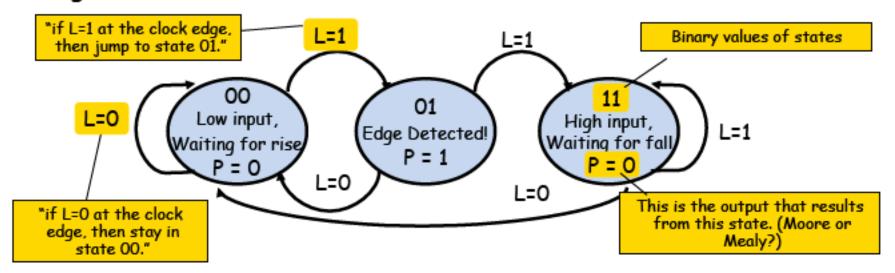


Step 1: State Transition Diagram

Block diagram of desired system:



 State transition diagram is a useful FSM representation and design aid:



In each state there should be one and only one arc for each input combination

Choosing State Representation

Choice #1: binary encoding

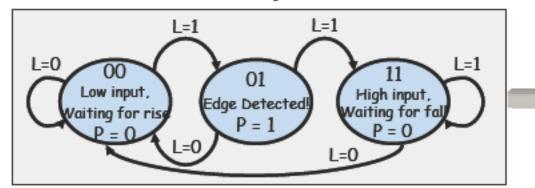
For N states, use ceil(log₂N) bits to encode the state with each state represented by a unique combination of the bits. Tradeoffs: most efficient use of state registers, but requires more complicated combinational logic to detect when in a particular state.

Choice #2: "one-hot" encoding

For N states, use N bits to encode the state where the bit corresponding to the current state is 1, all the others 0. Tradeoffs: more state registers, but often much less combinational logic since state decoding is trivial.

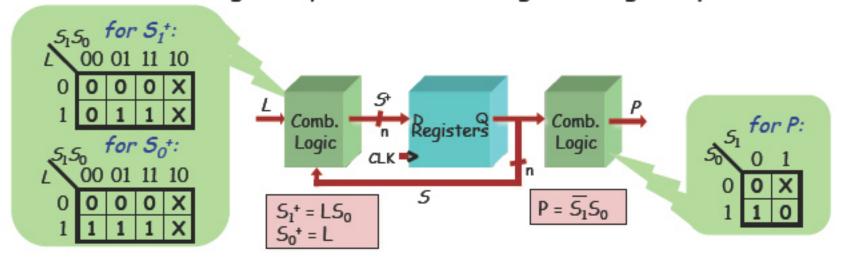
Step 2: Logic Derivation

Transition diagram is readily converted to a state transition table (just a truth table)

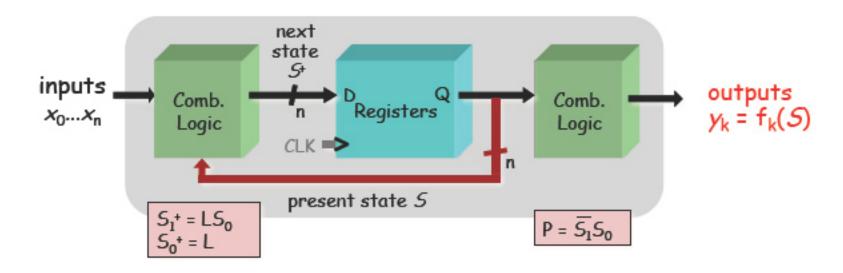


Current State		In	Next State		Out
5,	5 ₀	L	<i>5₁⁺</i>	5 ₀⁺	P
0	0	0	0	0	0
0	0	1	0	1	0
0	1	0	0	0	1
0	1	1	1	1	1
1	1	0	0	0	0
1	1	1	1	1	0

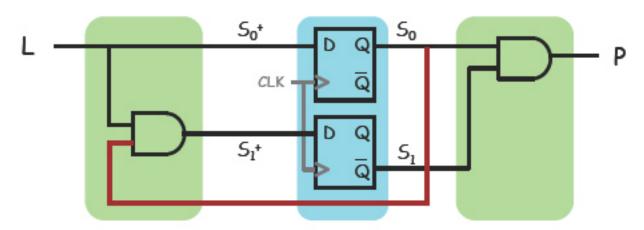
Combinational logic may be derived using Karnaugh maps



Moore Level-to-Pulse Converter



Moore FSM circuit implementation of level-to-pulse converter:





FSM Example

GOAL:

Build an electronic combination lock with a reset button, two number buttons (0 and 1), and an unlock output. The combination should be 01011.

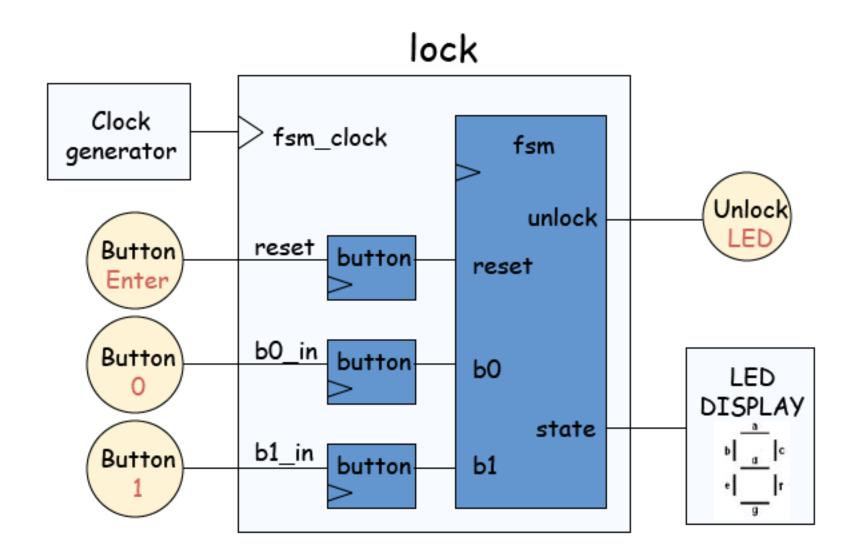


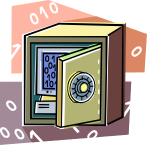
STEPS:

- Design lock FSM (block diagram, state transitions)
- Write Verilog module(s) for FSM

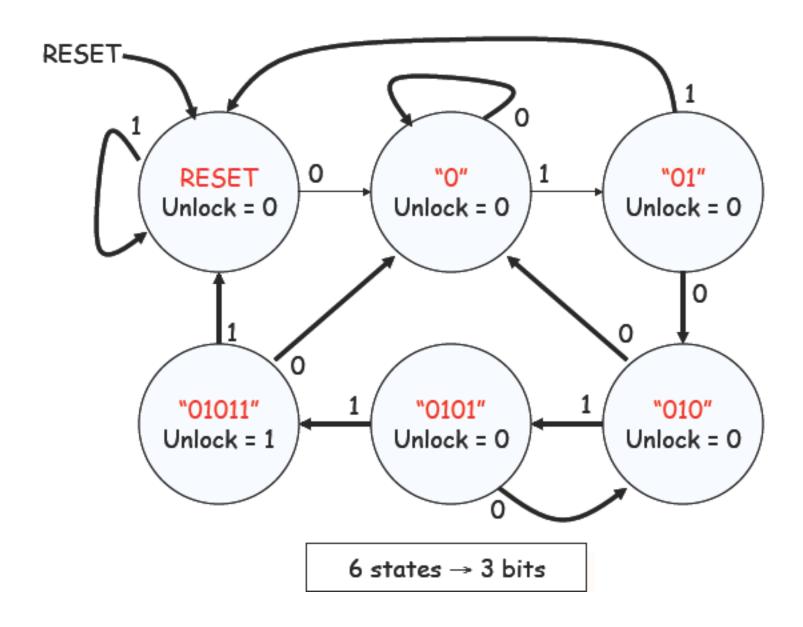


Step 1A: Block Diagram





Step 1B: State transition diagram





Step 2: Write Verilog

```
// synchronize push buttons, convert to pulses
  // implement state transition diagram
  reg [2:0] state,next_state;
  always @(*) begin
    // combinational logic!
    next_state = ???;
  end
  always @(posedge clk) state <= next_state;</pre>
  // generate output
  assign out = ???;
 // debugging?
endmodule.
```



Step 2A: Synchronize buttons

```
// button
// push button synchronizer and level-to-pulse converter
// OUT goes high for one cycle of CLK whenever IN makes a
// low-to-high transition.
module button(
  input clk, in,
  output out
);
                             clk.
  reg r1, r2, r3;
  always @(posedge clk)
                                   synchronizer
                                                  state
  begin
    r1 <= in; // first reg in synchronizer
    r2 <= r1; // second reg in synchronizer, output is in sync!
    r3 <= r2; // remembers previous state of button
  end
  // rising edge = old value is 0, new value is 1
  assign out = \simr3 & r2;
endmodule.
```



Step 2B: state transition diagram

```
parameter S_RESET = 0; // state assignments
parameter S_0 = 1:
parameter S_01 = 2;
parameter S_010 = 3;
                                               Unlock = 0
                                                       Unlock = 0
                                                               Unlock = 0
parameter S_0101 = 4;
parameter S_01011 = 5;
                                                                *010*
                                                *01011 *
                                                        "0101"
reg [2:0] state, next_state;
                                                               Unlock = 0
                                               Unlock = 1
                                                       Unlock = 0
always Q(*) begin
 // implement state transition diagram
 if (reset) next_state = S_RESET;
  else case (state)
    S_RESET: next_state = b0 ? S_0 : (b1 ? S_RESET : state);
    S_0: next_state = b0 ? S_0 : (b1 ? S_01 : state);
    S_01: next_state = b0 ? S_010 : (b1 ? S_RESET : state);
    S_010: next_state = b0 ? S_0 : (b1 ? S_0101 : state);
    S_0101: next_state = b0 ? S_010 : (b1 ? S_01011 : state);
    S_01011: next_state = b0 ? S_0 : (b1 ? S_RESET : state);
    default: next_state = S_RESET; // handle unused states
  endcase
end
always @(posedge clk) state <= next_state;
```



Step 2C: generate output

```
// it's a Moore machine! Output only depends on current state
assign out = (state == S_01011);
```

Step 2D: debugging?

```
// hmmm. What would be useful to know? Current state?
assign hex_display = {1'b0,state[2:0]};
```



Step 2: final Verilog implementation

```
module lock(input clk, reset in, b0 in, b1 in,
            output out, output [3:0] hex display);
  wire reset, b0, b1; // synchronize push buttons, convert to pulses
  button b reset(clk, reset in, reset);
  button b 0(clk,b0 in,b0);
  button b 1(clk,bl in,bl);
  parameter S RESET = 0; parameter S 0 = 1; // state assignments
  parameter S 01 = 2; parameter S 010 = 3;
  parameter S 0101 = 4; parameter S 01011 = 5;
  reg [2:0] state, next state;
  always @(*) begin
                                        // implement state transition diagram
    if (reset) next state = S RESET;
    else case (state)
      S_RESET: next_state = b0 ? S_0 : (b1 ? S_RESET : state);
      S_0: next_state = b0 ? S_0 : (b1 ? S_01 : state);
S_01: next_state = b0 ? S_010 : (b1 ? S_RESET : state);
      S 010: next state = b0 ? S 0 : (b1 ? S 0101 : state);
      S 0101: next state = b0 ? S 010 : (b1 ? S 01011 : state);
      S_01011: next_state = b0 ? S_0 : (b1 ? S_RESET : state);
      default: next state = S RESET; // handle unused states
    endcase
  end
  always @(posedge clk) state <= next state;
  assign out = (state == S_01011);  // assign output: Moore machine
  assign hex display = {l'b0,state}; // debugging
endmodule
```

Partitioning state machines

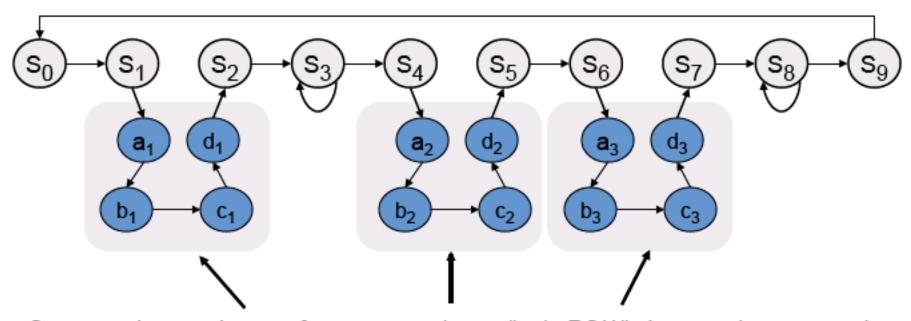
Basically the trick is that you need to put the top one on hold

Partitioning state machines

Basically the trick is that you need to put the top one on hold

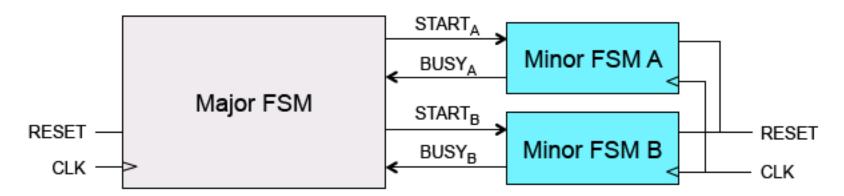
Toward FSM Modularity

Consider the following abstract FSM:



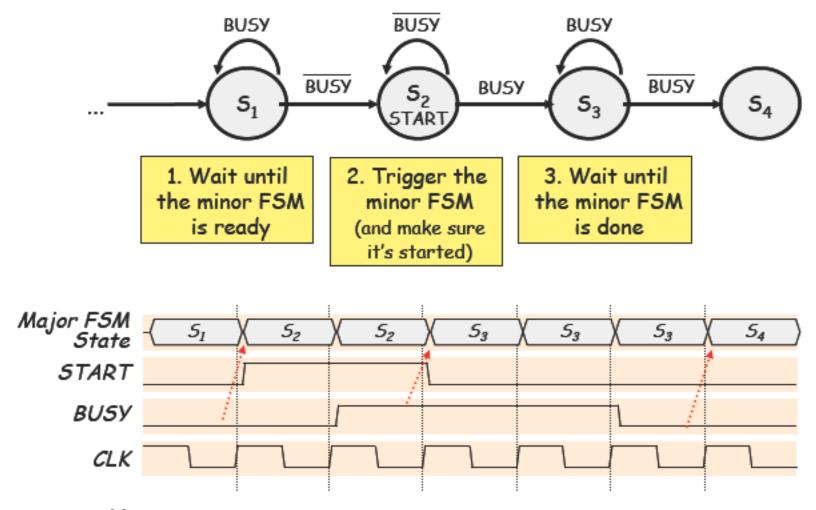
- Suppose that each set of states a_x...d_x is a "sub-FSM" that produces exactly the same outputs.
- Can we simplify the FSM by removing equivalent states?
 No! The outputs may be the same, but the next-state transitions are not.
- This situation closely resembles a procedure call or function call in software...how can we apply this concept to FSMs?

The Major/Minor FSM Abstraction



- Subtasks are encapsulated in minor FSMs with common reset and clock
- Simple communication abstraction:
 - START: tells the minor FSM to begin operation (the call)
 - BUSY: tells the major FSM whether the minor is done (the return)
- The major/minor abstraction is great for...
 - Modular designs (always a good thing)
 - Tasks that occur often but in different contexts
 - Tasks that require a variable/unknown period of time
 - Event-driven systems

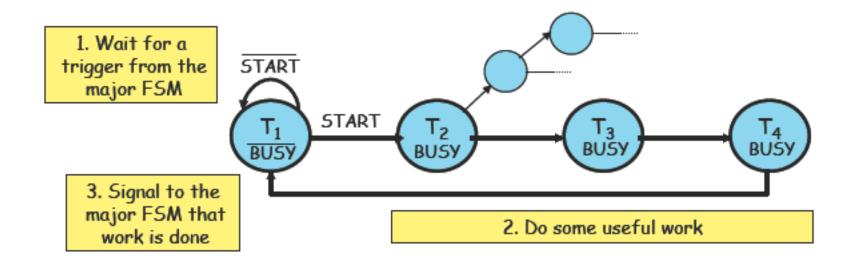
Inside the Major FSM

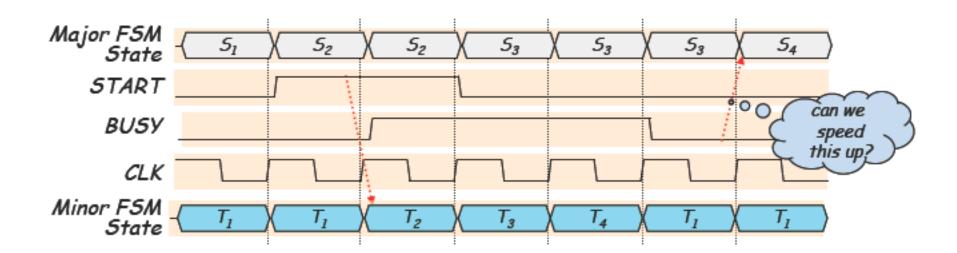


Variations:

- Usually don't need both Step 1 and Step 3
- One cycle "done" signal instead of multi-cycle "busy"

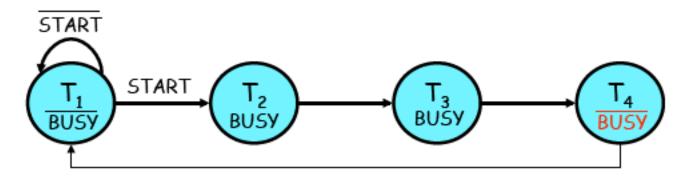
Inside the Minor FSM

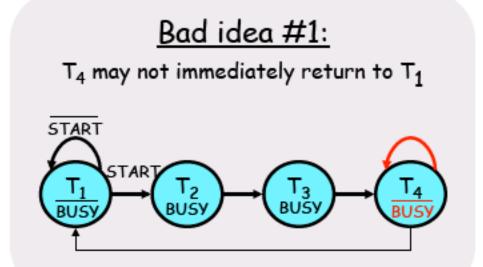


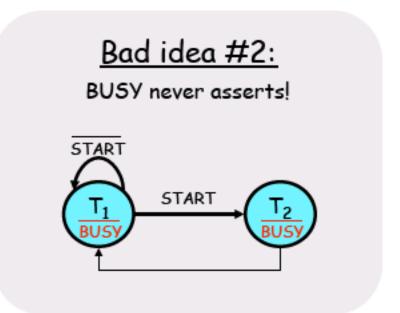


Optimizing the Minor FSM

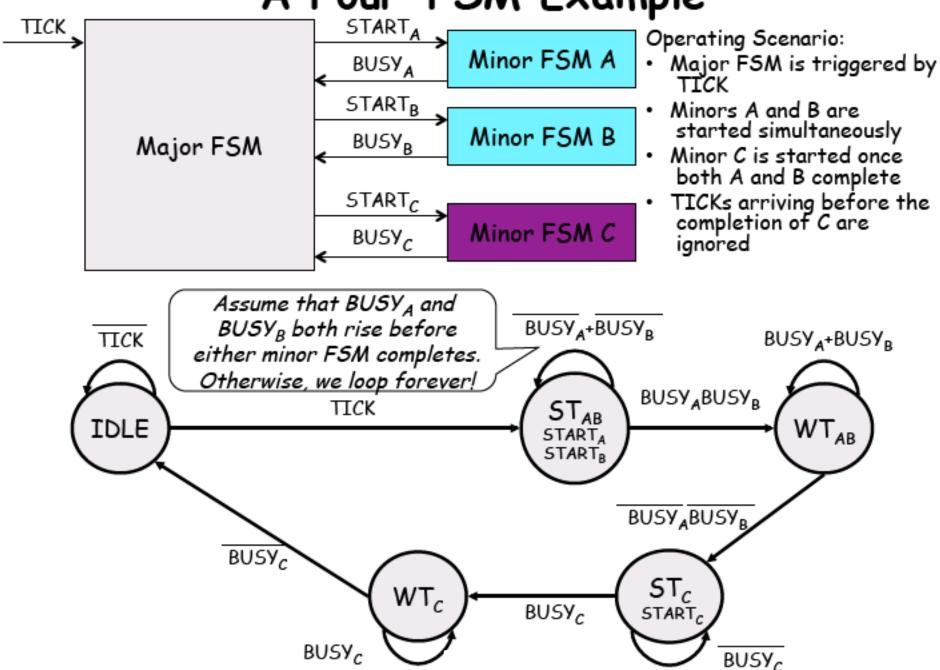
Good idea: de-assert BUSY one cycle early



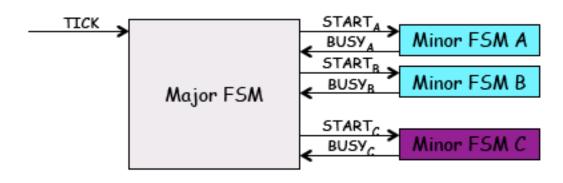


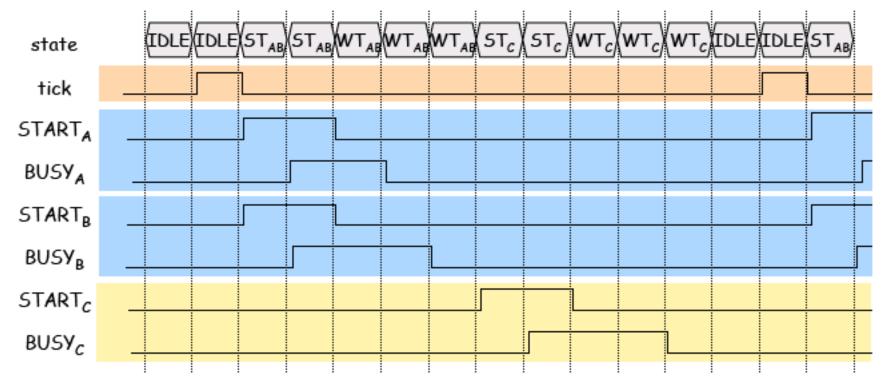


A Four-FSM Example



Four-FSM Sample Waveform





How to write a state machine in verilog

```
input k1,k2,k3, c1k;
output 11,12,13;
//create registers
//to "store state"
reg [3:0] state, next;
//use parameters to give names
//to states
parameter s1=0;
parameter s2=1;
parameter s3=2:
parameter s4=3:
//The core code
//this, and only this,
//changes the states
always @ (posedge clk)
       state<=next;</pre>
```

```
//decide which will be
//the next state
always @ (state or k1 or k2 or k3)
 case (state)
  S1:
     if(????) next=????;
     else next=S1;
  S2:
     if(????) next=????;
     else next=S2; //The wait modes
  S3:
     if(????) next=????;
     else next=S3;
//compute output
//according to state and/or input
  assign 11= S1;
  assign 12=S2 || S3;
  assign 13=S3;
```

A vending machine

- All selections are \$0.30.
- The machine makes change.
 (Dimes and nickels only.)
- Inputs: limit 1 per clock
 - □ Q quarter inserted
 - □ D dime inserted
 - N nickel inserted
- Outputs: limit 1 per clock
 - □ DC dispense can
 - □ DD dispense dime
 - □ DN dispense nickel



What states?

A starting (idle) state:



A state for each possible amount of money captured:



What's the maximum amount of money captured before purchase?
25 cents (just shy of a purchase) + one quarter (largest coin)



States to dispense change (one per coin dispensed):

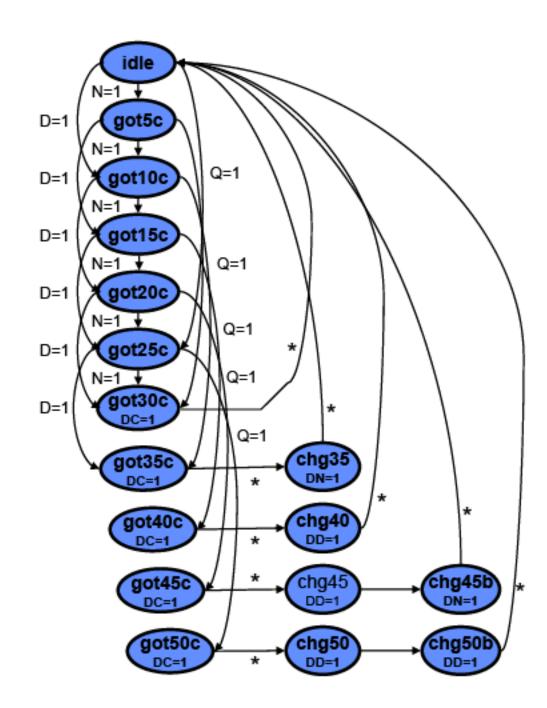




A Moore vender

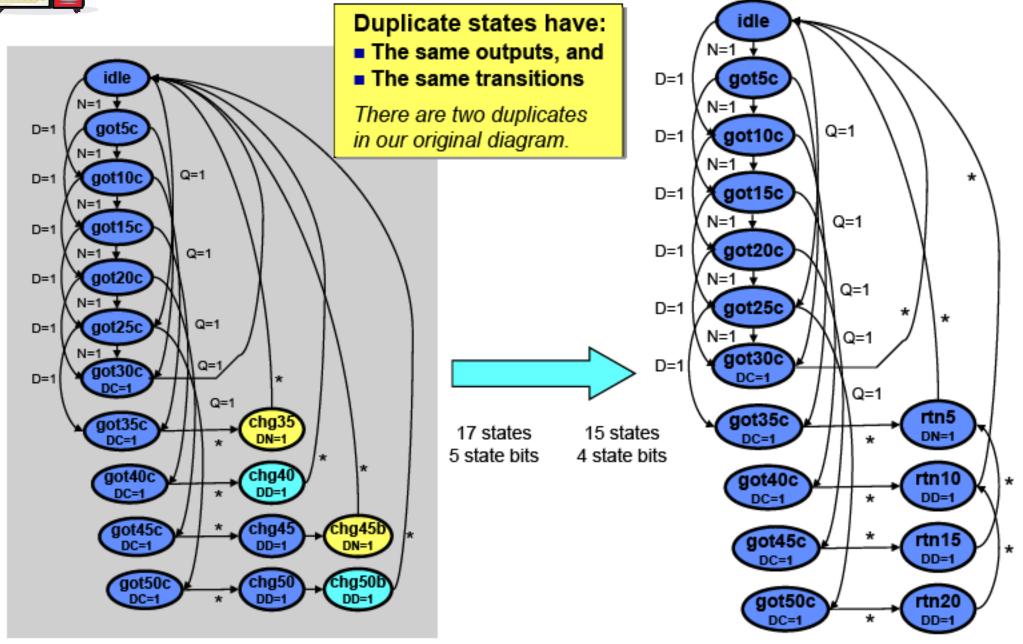
Here's a first cut at the state transition diagram.





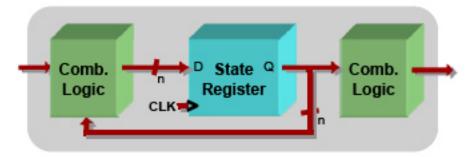


State reduction





The verilog code



FSMs are easy in Verilog. Simply write one of each:

- State register (sequential always block)
- Next-state combinational logic (comb. always block with case)
- Output combinational logic block (comb. always block or assign statements)

```
module mooreVender (N, D, Q, DC, DN, DD, clk, reset, state);
input N, D, Q, clk, reset;
output DC, DN, DD;
output [3:0] state;
reg [3:0] state, next;
```

States defined with parameter keyword

```
parameter IDLE = 0;
parameter GOT_5C = 1;
parameter GOT_10C = 2;
parameter GOT_15C = 3;
parameter GOT_20C = 4;
parameter GOT_25C = 5;
parameter GOT_30C = 6;
parameter GOT_35C = 7;
parameter GOT_40C = 8;
parameter GOT_45C = 9;
parameter GOT_50C = 10;
parameter RETURN_20C = 11;
parameter RETURN_15C = 12;
parameter RETURN_15C = 13;
parameter RETURN_5C = 14;
```

State register defined with sequential always block

```
always @ (posedge clk or negedge reset)
if (!reset) state <= IDLE;
else state <= next;</pre>
```



The Verilog code

Next-state logic within a combinational always block

```
always @ (state or N or D or Q) begin
   case (state)
     IDLE:
                if (Q) next = GOT 25c;
           else if (D) next = GOT 10c;
              else if (N) next = GOT 5c;
              else next = IDLE;
               if (Q) next = GOT_30c;
     GOT 5c:
           else if (D) next = GOT_15c;
              else if (N) next = GOT 10c;
             else next = GOT 5c;
     GOT_10c: if (Q) next = GOT_35c;
           else if (D) next = GOT_20c;
              else if (N) next = GOT 15c;
             else next = GOT 10c;
     GOT_15c: if (Q) next = GOT_40c;
           else if (D) next = GOT 25c;
              else if (N) next = GOT 20c;
              else next = GOT 15c;
     GOT 20c: if (Q) next = GOT 45c;
           else if (D) next = GOT 30c;
              else if (N) next = GOT 25c;
             else next = GOT 20c;
```

```
GOT 25c:
          if (Q) next = GOT 50c;
          else if (D) next = GOT 35c;
             else if (N) next = GOT_30c;
              else next = GOT 25c;
     GOT 30c: next = IDLE;
     GOT 35c: next = RETURN 5c;
     GOT 40c: next = RETURN 10c;
     GOT 45c: next = RETURN 15c;
     GOT 50c: next = RETURN 20c;
     RETURN 20c: next = RETURN 10c;
     RETURN 15c: next = RETURN 5c;
     RETURN 10c: next = IDLE;
     RETURN 5c:
                 next = IDLE;
     default: next = IDLE;
   endcase
 end
```

Combinational output assignment







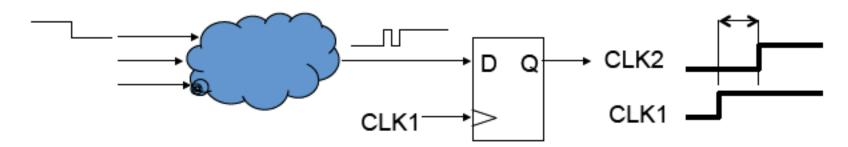


Where should CLK come from?

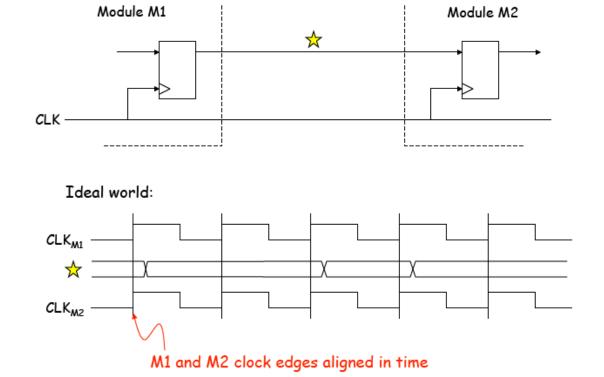
- Option 1: external crystal
 - Stable, known frequency, typically 50% duty cycle
- Option 2: internal signals
 - Option 2A: output of combinational logic



- No! If inputs to logic change, output may make several transitions before settling to final value → several rising edges, not just one! Hard to design away output glitches...
- Option 2B: output of a register
 - Okay, but timing of CLK2 won't line up with CLK1

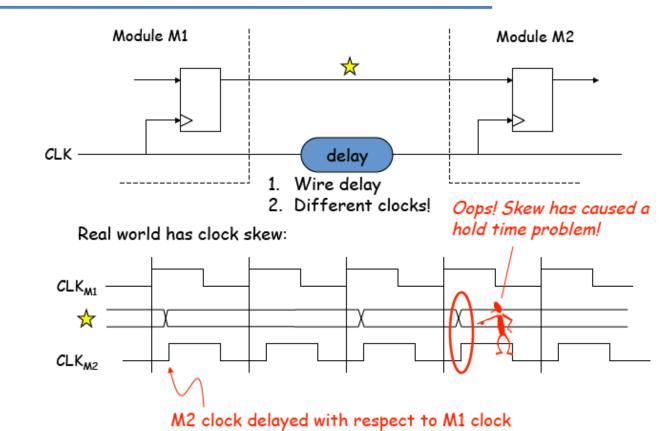


Clocking in and ideal world



Clocking in a not so perfect reality

FPGAs contain global clock lines that have low skew. Number is limited



Clocks with periods multiple of two

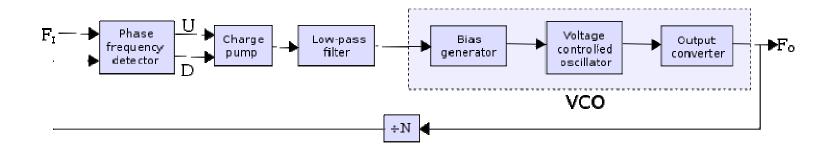
```
reg clk2,clk4,clk8,clk16;
always @(posedge clk) clk2 <= ~clk2;
always @(posedge clk2) clk4 <= ~clk4;
always @(posedge clk4) clk8 <= ~clk16;
always @(posedge clk8) clk16 <= ~clk16;
```

Solution: 1 clock, many enables

Use one (high speed) clock, but create enable signals to select a subset of the edges to use for a particular piece of sequential logic

```
reg [3:0] count;
  always @(posedge clk) count <= count + 1; // counts 0..15
  wire enb2 = (count[0] == 1'b1);
  wire enb4 = (count[1:0] == 2'b11);
                                                always @(posedge clk)
  wire enb8 = (count[2:0] == 3'b111);
                                                 if (enb2) begin
                                                   // get here every 2<sup>nd</sup> cycle
  wire enb16 = (count[3:0] == 4'b1111);
                                                  end
CLK
     14
         > 15
                                                        10
                                                            11
                                                                    13
                                                                        14
count
ENB2
ENB4
ENB8
ENB16
```

The Phase Locked Loop (PLL)





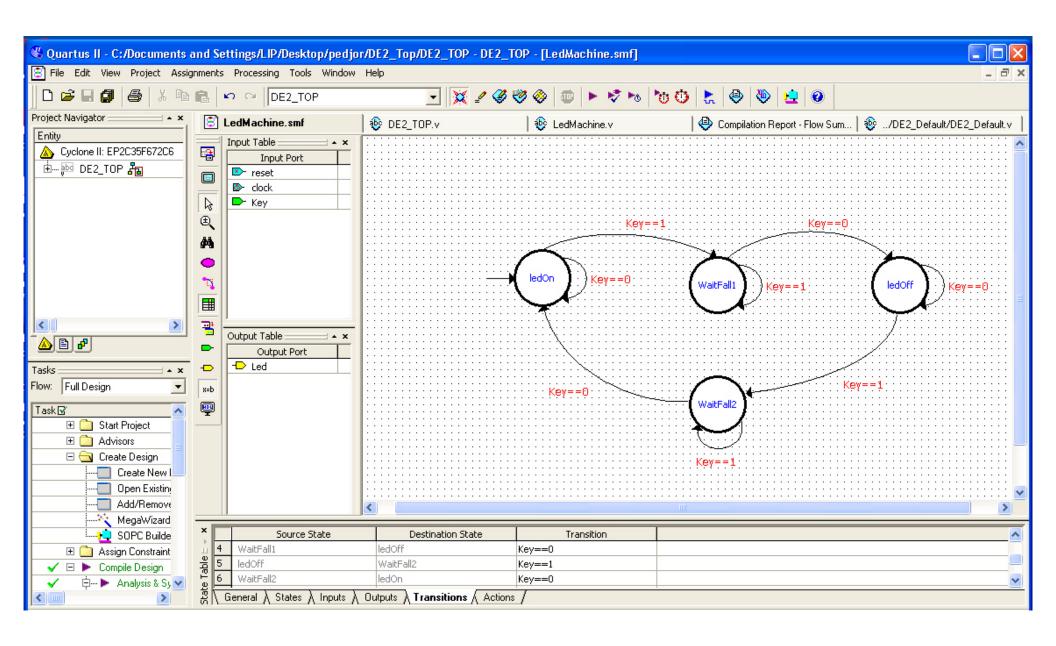
A feature available in Cyclone II devices that employs a phase-locked loop (PLL). The Cyclone II PLL provides advanced multiplication, programmable duty cycle, phase shifting, programmable bandwidth, manual clock switchover, clock outputs driving all networks, and a source synchronous mode.

You can take advantage of the Cyclone II PLL with the altpll megafunction.

Phase-Locked Loops (ALTPLL)

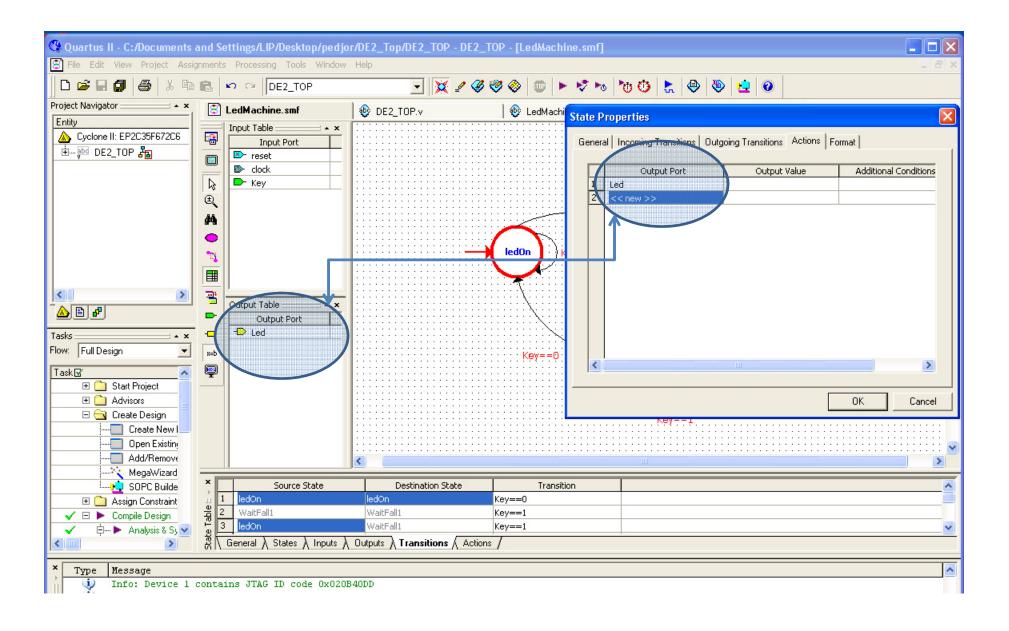
Megafunction User Guide

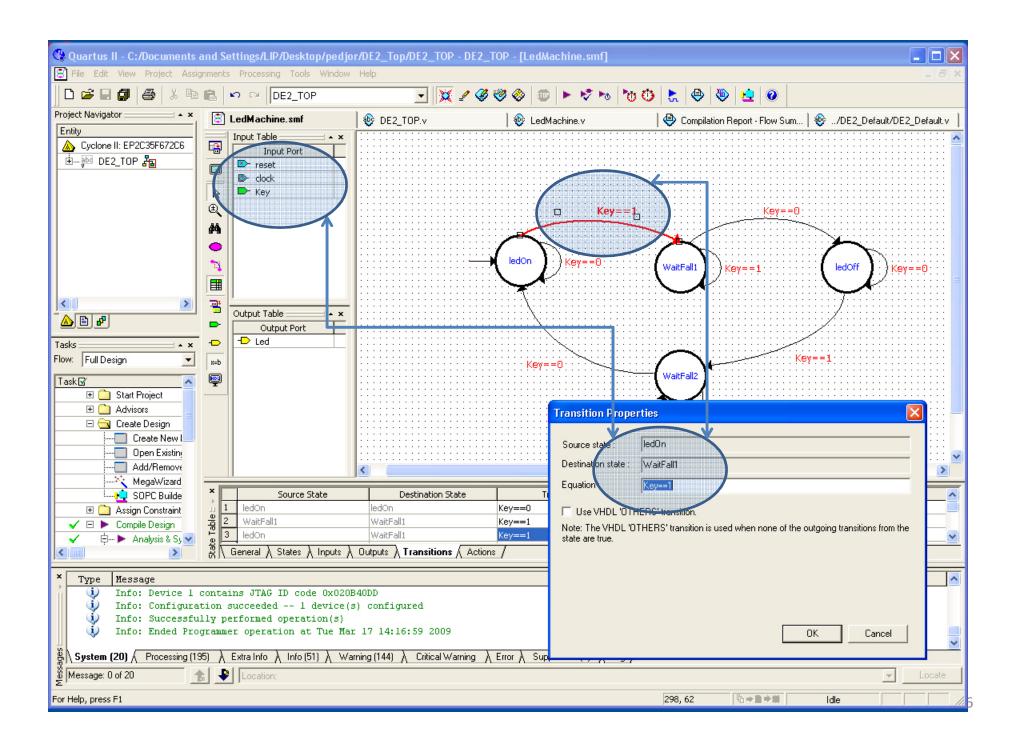
File > new > statemachine

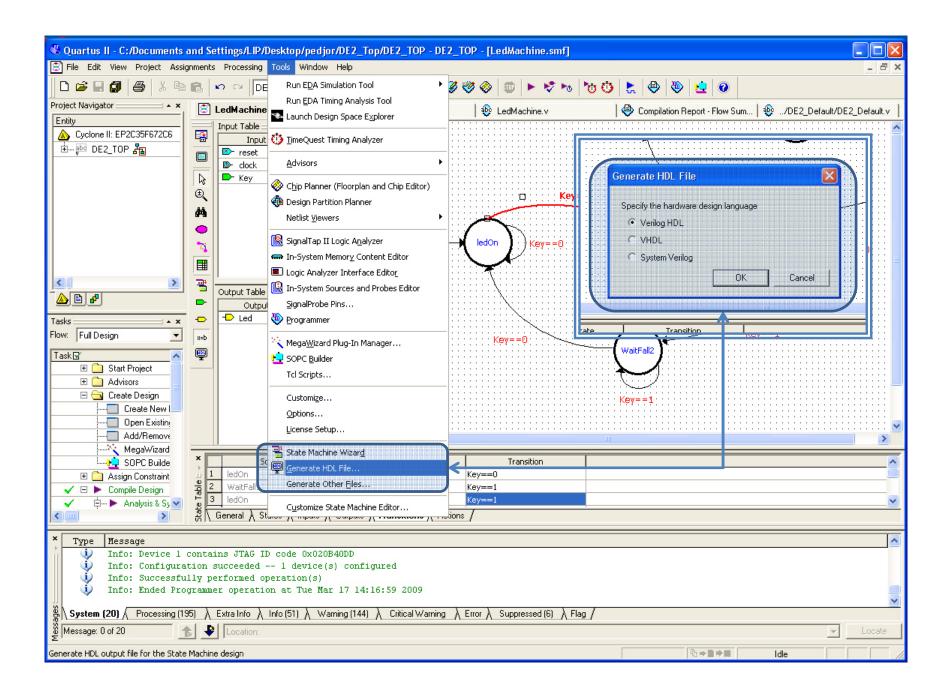


For more information about creating and editing state machine diagrams, refer to the Quartus II Help.

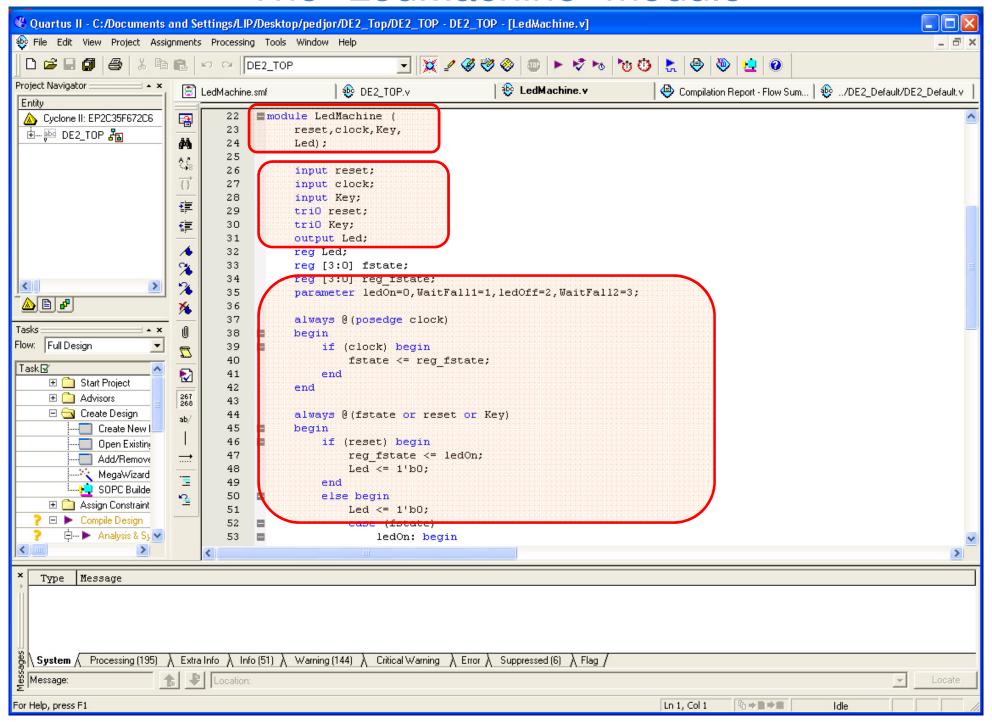
Define the states output







The "LedMachine" module



The top design

